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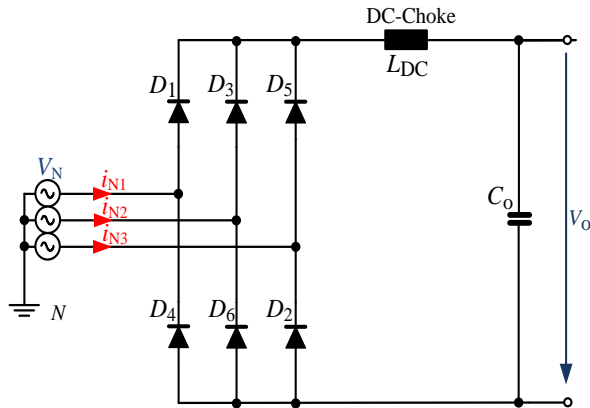
Active Three - Phase Rectifier System Using a Flying Converter Cell

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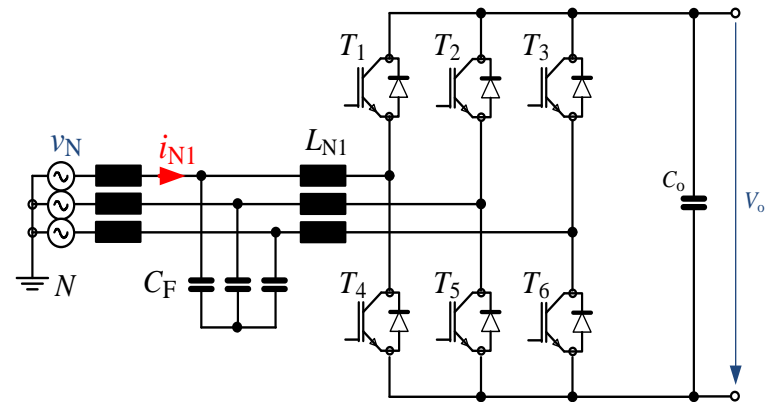
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Passive Rectifier Circuit



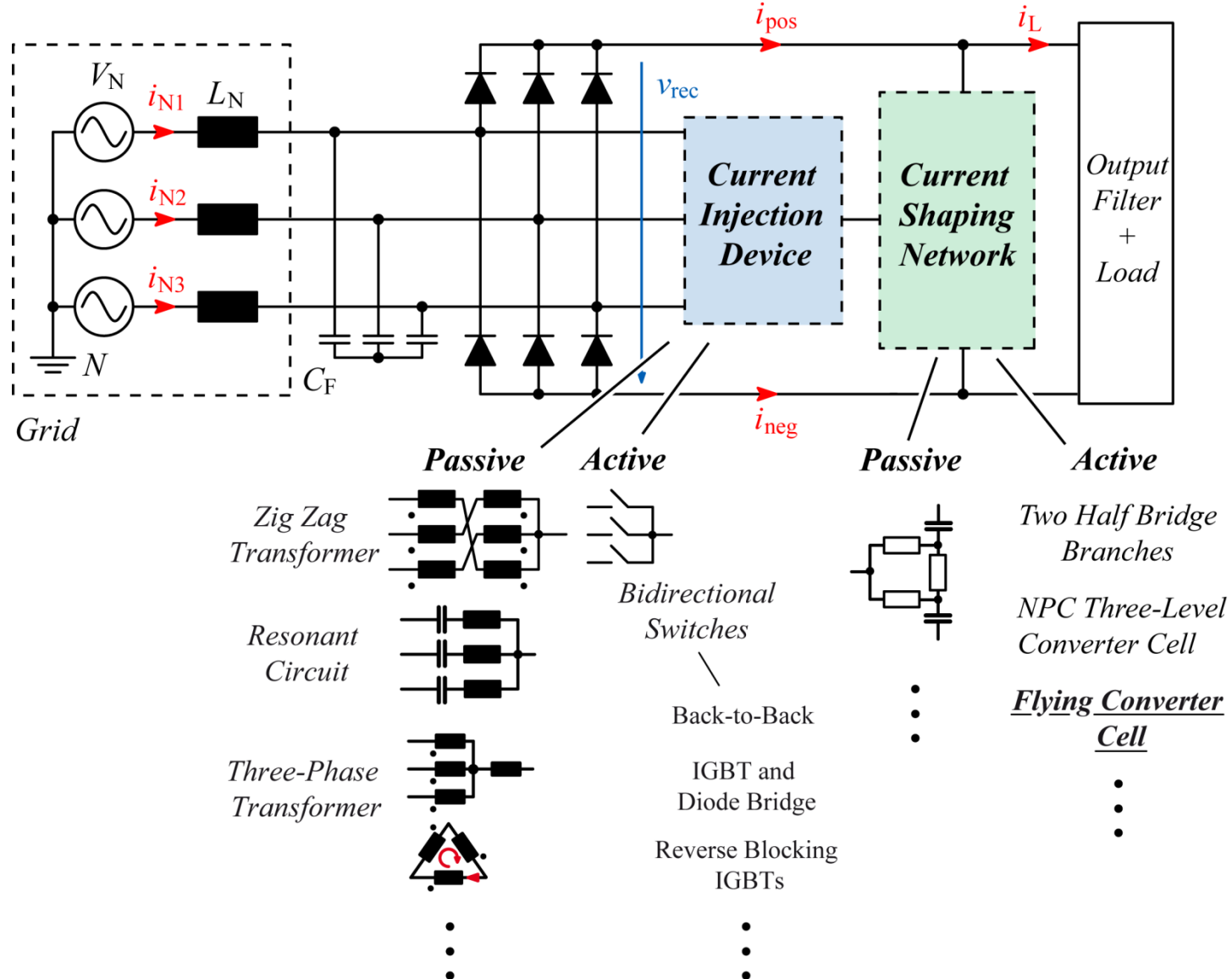
- Widely used in Industry
- Simple and Robust
- High Efficiency
- Poor Input Current Quality
 $THD_i=48\%$
- Power Factor of 0.9...0.95
- No Active Output Voltage Control

Active 2-Level Rectifier Circuit

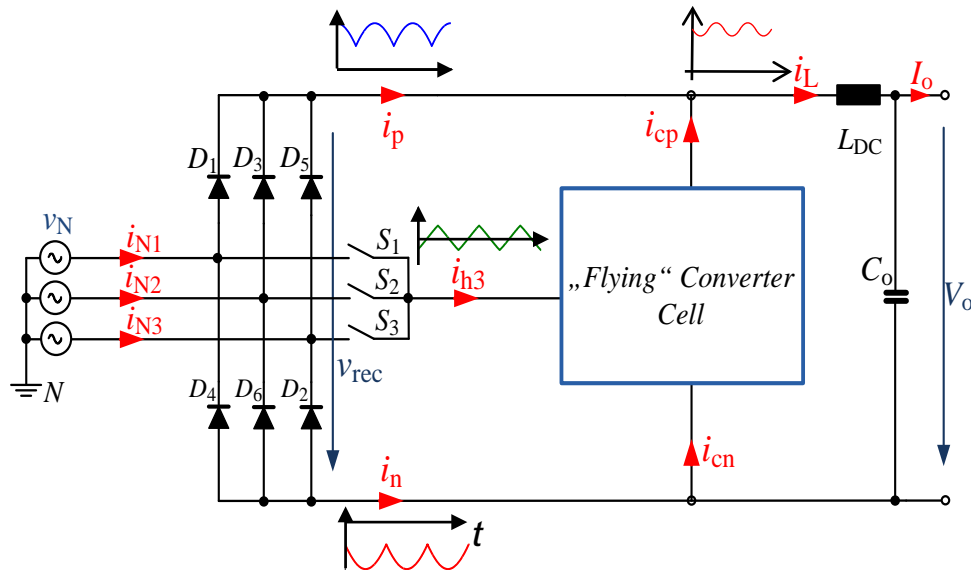


- Industry Standard
- Input Current Quality $THD_i < 5\%$
- High Power Factor
- Controlled Output Voltage
- Active Switches have to Process full Output Power → **Reduced Efficiency**
- **Existing Passive Rectifier Bridges cannot be Extended to the Active 2-Level Rectifier Topology.**

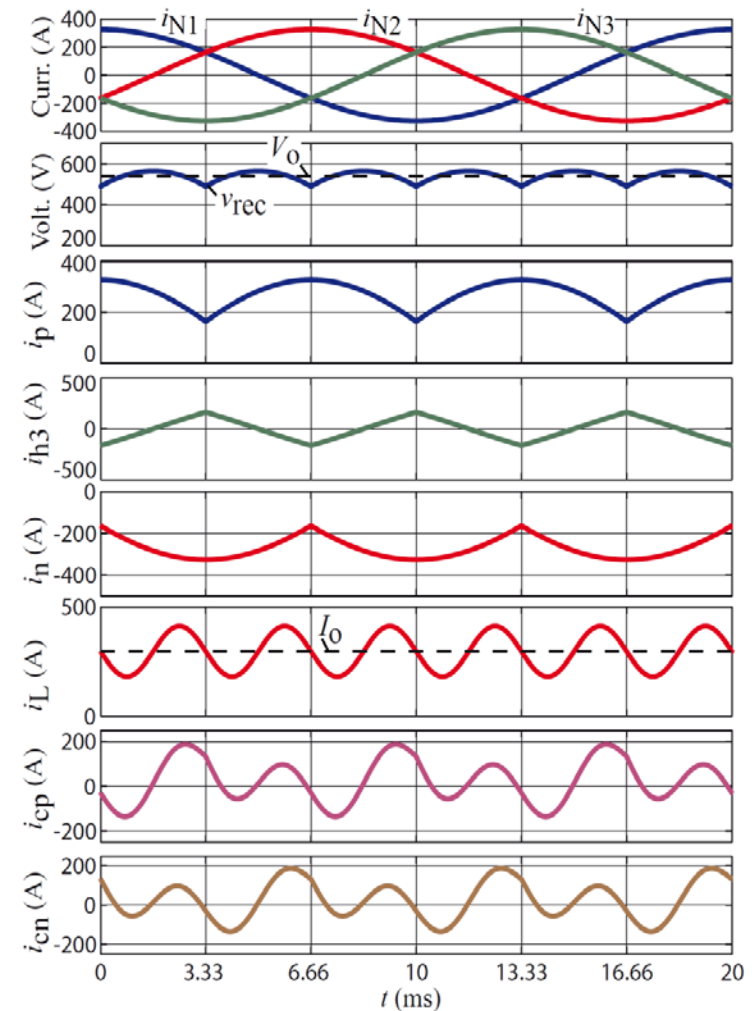
Third Harmonic Injection

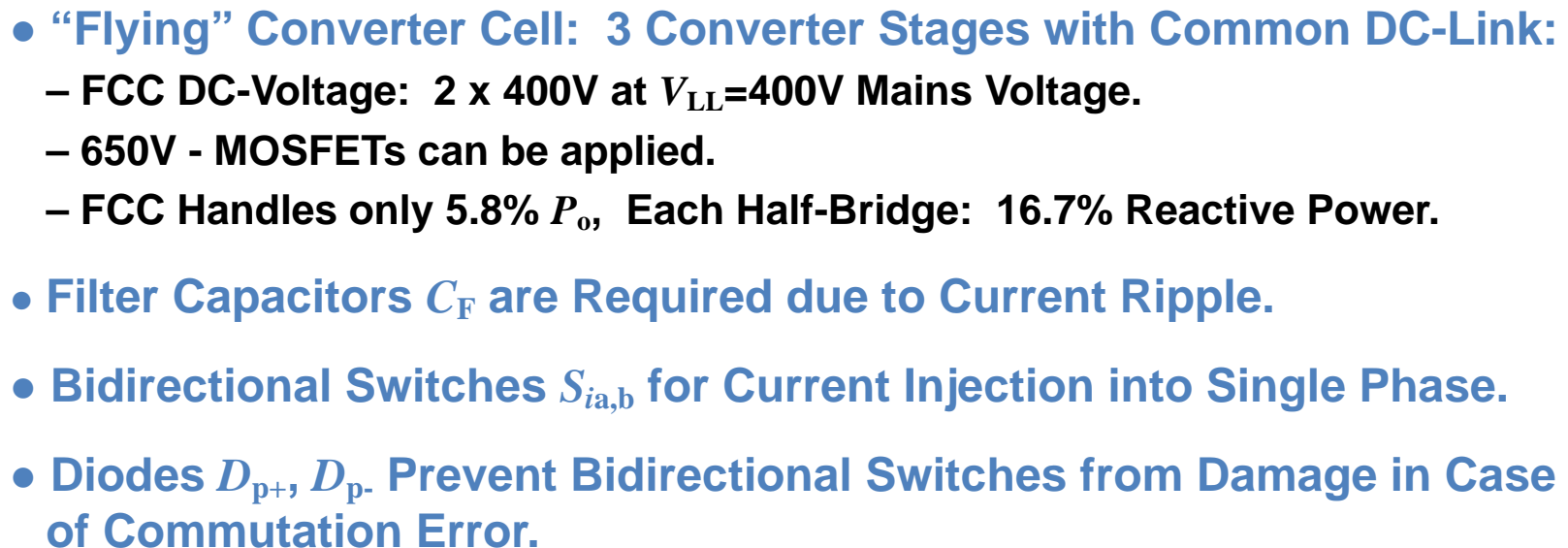


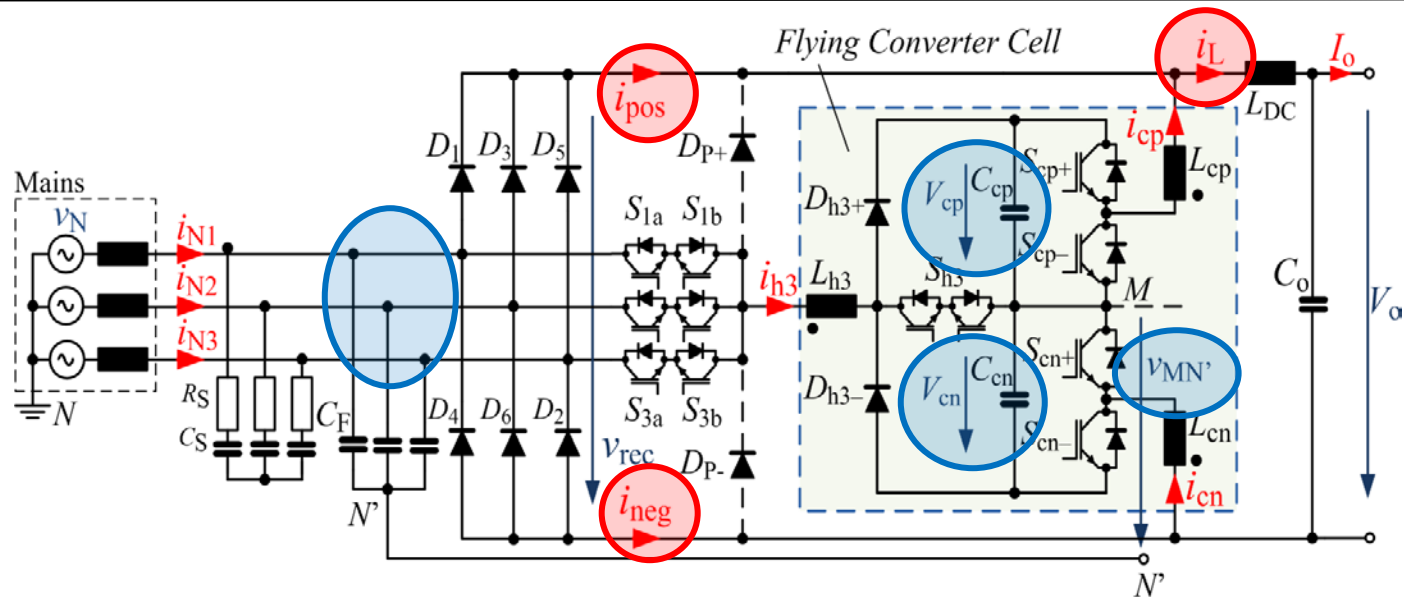
“Flying” Converter Cell – Basic Concept



- **Extension of Existing Diode Bridge by Inserting a “Flying” Converter Cell**
Injects Current into :
 - One Phase of the Diode-Bridge (i_{h3}).
 - Positive / negative Bus-Bar (i_{cp} , i_{cn}).
- **No Active Control of Output Voltage.**
 - Output Voltage is Defined by Mains.
- **Constant Output Voltage:**
 - Output Capacitor C_o can be Considerably Large.
 - No High-Frequency Common Mode Voltage at the DC-Link!







- **Required Controllers:**

- Current Controller to Achieve Sinusoidal Input Currents.
- Control of Total FCC DC-Voltage.
- Balancing of FCC DC-Voltages.

- **Required Measurement Signals:**

- Currents: i_{pos} , i_{neg} , i_L
- Voltages: $v_{N1} \dots v_{N3}$, $v_{MN'}$, V_{cp} , V_{cn} , v_{rec}

- Average mode current control.

- System equations:

$$\delta_{cp} v_{cp} + v_{MN,avg} - v_{pos} = L \frac{di_{cp}}{dt}$$

$$(1 - \delta_{h3}) (-v_{cn}) + v_{MN,avg} - v_{mid} = -L \frac{di_{h3}}{dt}$$

$$(1 - \delta_{cn}) (-v_{cn}) + v_{MN,avg} - v_{neg} = -L \frac{di_{cn}}{dt}$$

- Idea is to control i_{pos} and i_{neg}

- i_L is disturbance input.

- Current i_{h3} is defined by: $i_{cp} = i_{cn} + i_{h3}$

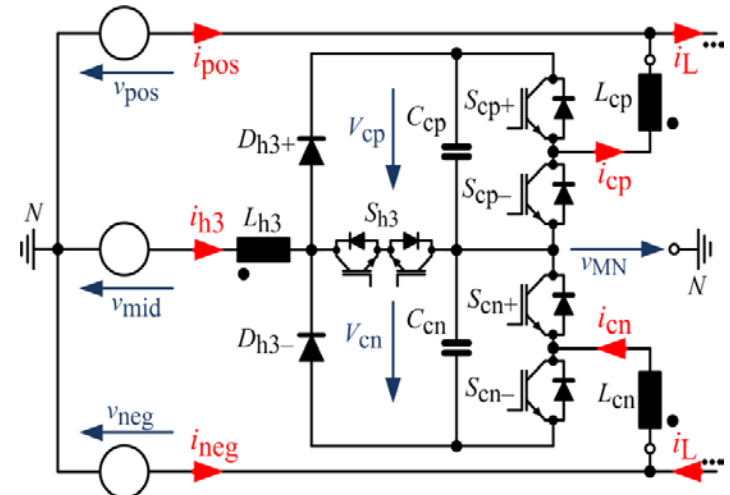
- Using $i_{pos} = i_L - i_{cp}$ and Applying Laplace Transformation:

$$sLi_{pos} = sLi_L - \delta_{cp}V_c - V_{MN,avg} + v_{pos}$$

- Feed-Forward Signal:

$$\delta_{cp} = \tilde{\delta}_{cp} + \frac{v_{pos} - V_{MN,avg}}{V_c} + s \frac{Li_L}{V_c}$$

Model of Converter

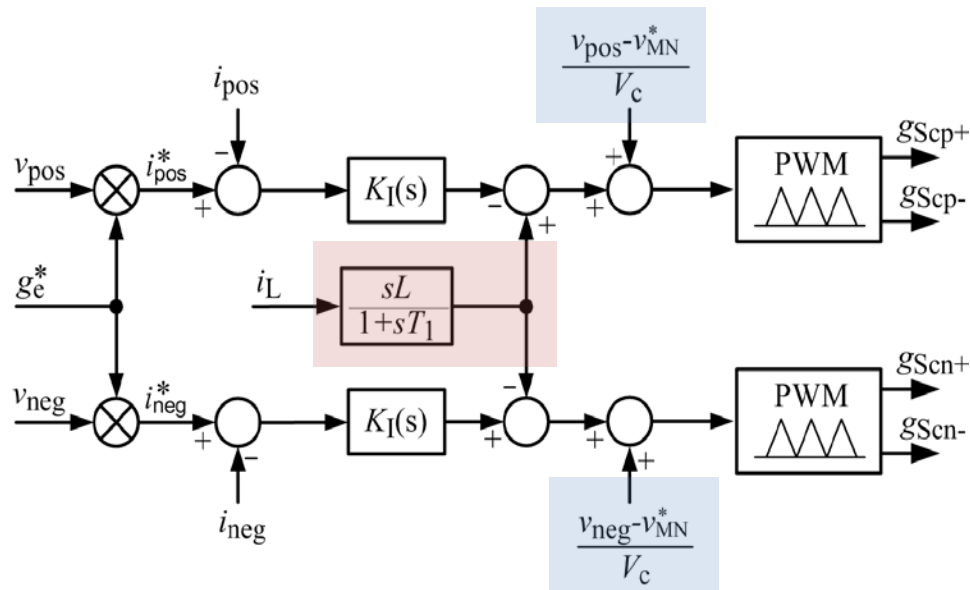


$$\varphi_N \in [0 \dots \frac{\pi}{6}] \quad (i_{N1}(\varphi_N) > 0; i_{N2}(\varphi_N), i_{N3}(\varphi_N) < 0)$$

- Simple Model:

$$G_I(s) = \frac{i_{pos}(s)}{\tilde{\delta}_{cp}(s)} = -\frac{V_c}{sL}$$

- Structure of Current Controller:



$$\delta_{cp} = \tilde{\delta}_{cp} + \frac{v_{pos} - V_{MN,avg}}{V_c} + s \frac{Li_L}{V_c}$$

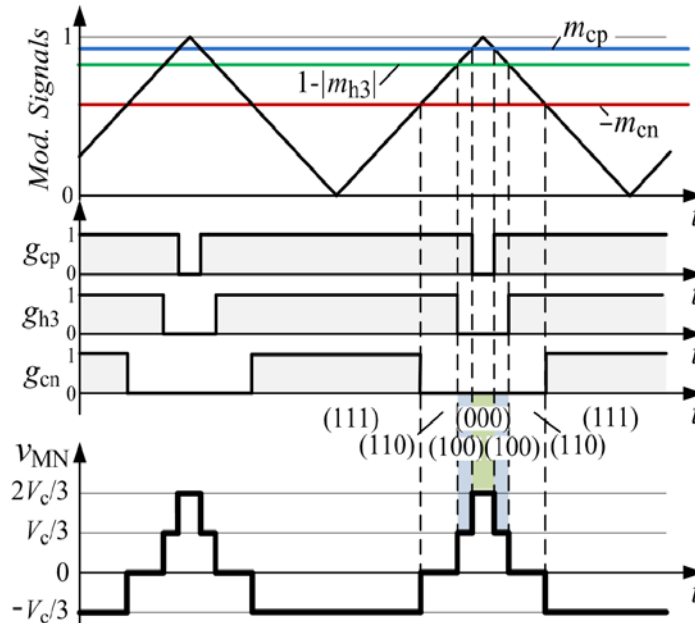
- Generation of Reference Currents using equivalent Conductance:

$$g_e^* = \frac{P_{in}}{V_{N1,rms}^2 + V_{N2,rms}^2 + V_{N3,rms}^2}$$

- Measurement of Power Transferred to Load:

- At the Input: $p(t) = v_{N1}(t)i_{N1}(t) + v_{N2}(t)i_{N2}(t) + v_{N3}(t)i_{N3}(t)$
- After Rectifier Diodes: $p(t) = v_{rec}(t)i_L(t)$

Carrier Signals in Phase



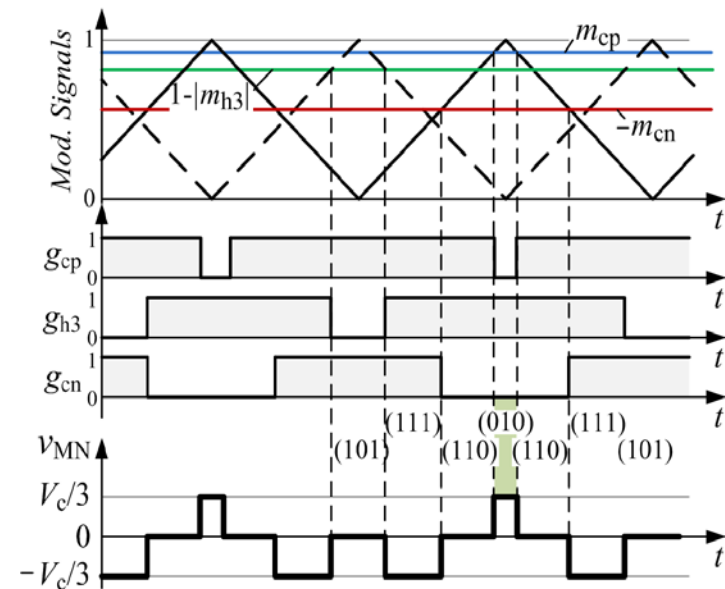
- **Higher FCC Mid-Point Voltage.**
- **Reduced Current Ripple:**

$$\Delta i_{c_n^p, \text{pkpk}} = \frac{V_c M^2}{4 f_s L}$$

$$\Delta i_{h3, \text{pkpk}} = \frac{V_c \frac{M}{2} \left(1 - \frac{M}{2}\right)}{f_s L}$$

1 : 1.5!

Carrier Signals 180° out of Phase

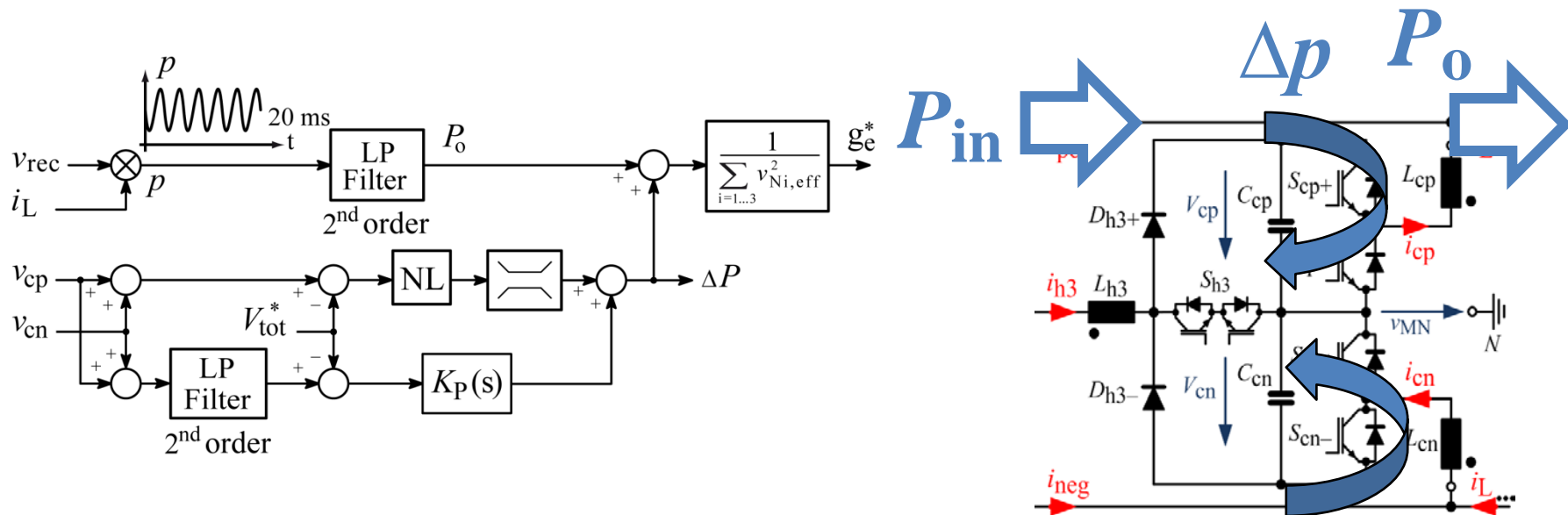


- **Reduced FCC Mid-Point Voltage.**
- **Increased Current Ripple:**

$$\Delta i_{c_n^p, \text{pkpk}} = \frac{V_c \left(M \frac{\sqrt{3}}{2} - \frac{1}{3} \right) \left(1 - M \frac{\sqrt{3}}{2} \right)}{f_s L_c}$$

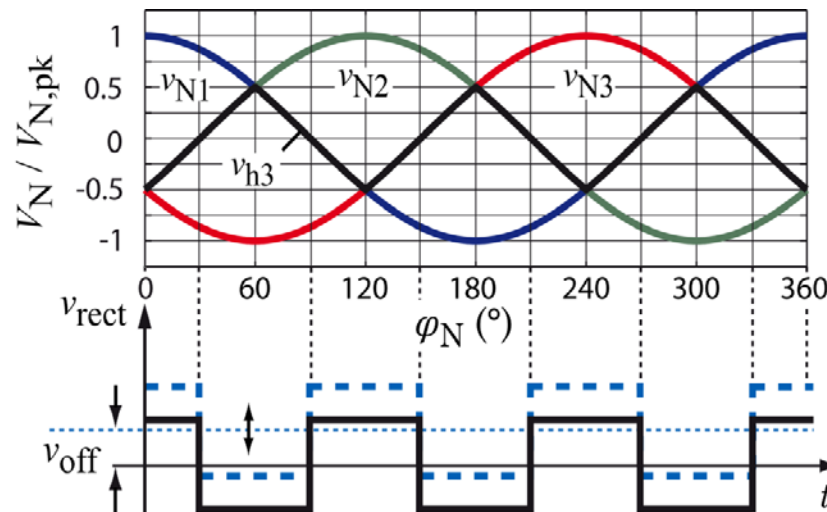
$$\Delta i_{h3, \text{pkpk}} = \frac{V_c \frac{1}{3} \left(1 - M \frac{\sqrt{3}}{2} \right)}{f_s L_c}$$

- **Control of Total FCC DC-Voltage by Voltage Controller:**
 - **Without Disturbing Mains Current.**
- **Balancing of FCC DC-Voltages by Dedicated Controller.**

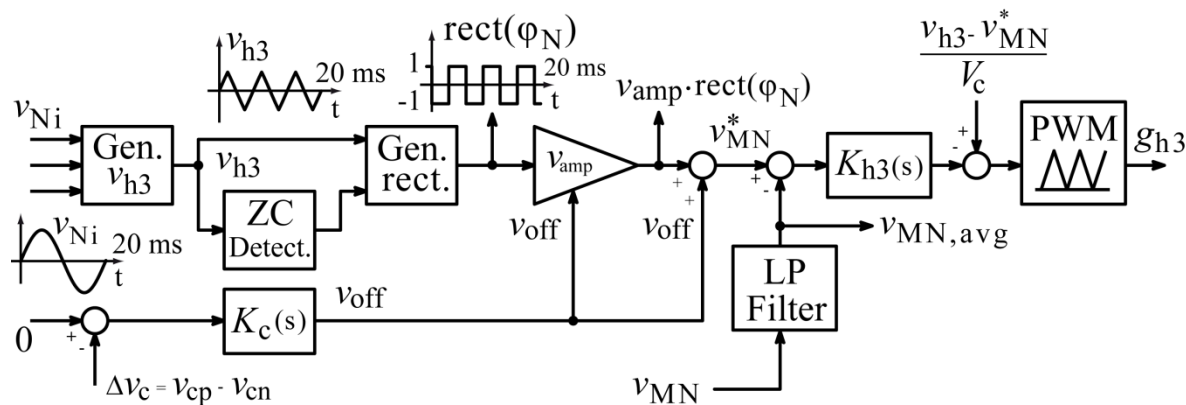


- **Control by Increasing/Reducing the Input Power:**
 - **Remaining Power is Transferred into FCC Capacitors.**

- Different Mid-Point Voltages v_{MN} Show Different Currents in V_{cp} and V_{cn} .
- Mid-Point Voltage Variations are Used for Balancing.
- Without Disturbing Input Currents.



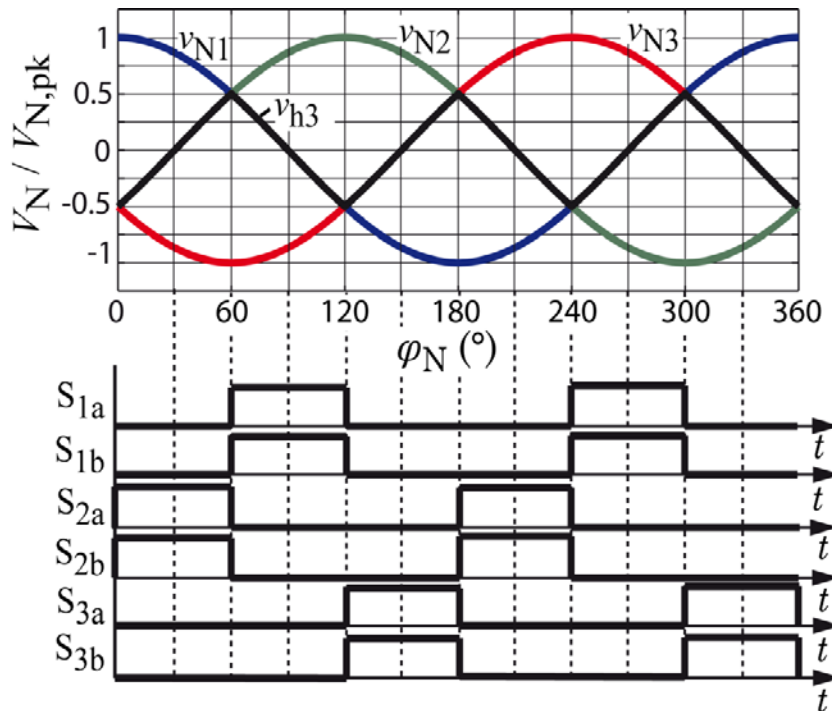
- Rectangular Shaped Mid-point Voltage Variations.
- Capacitors are Charged / Discharged Depending on the Offset v_{off} .



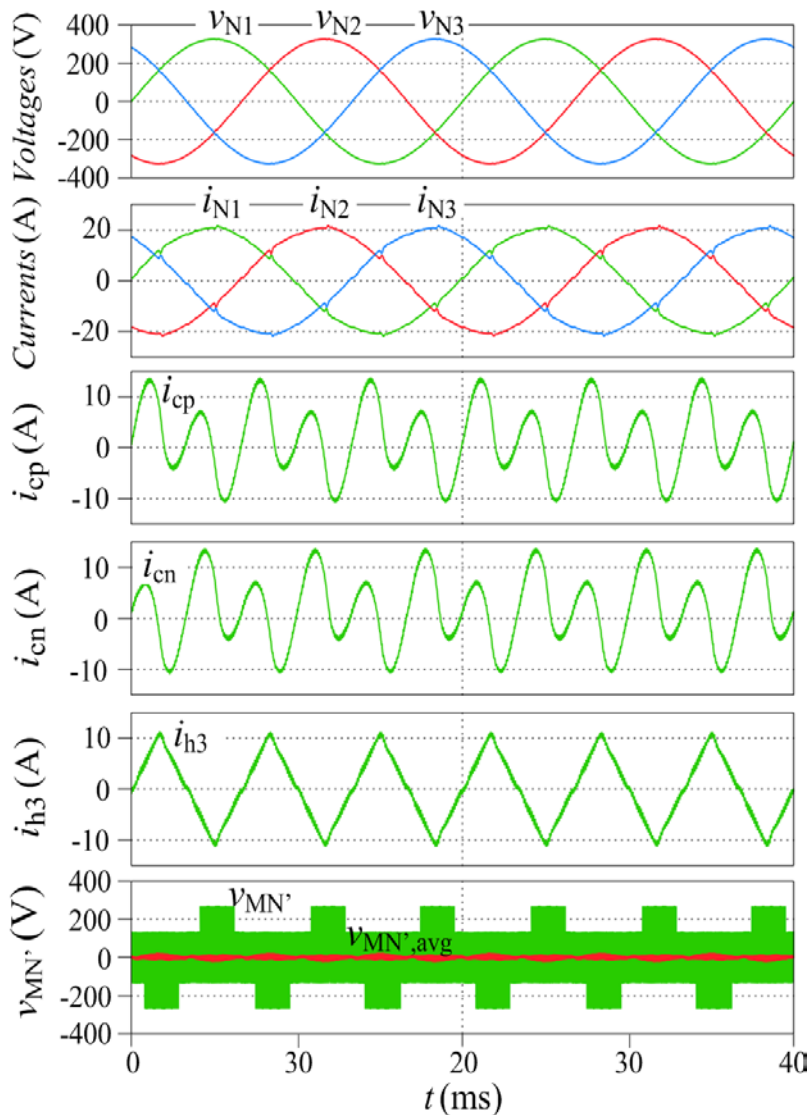
- Three-Level Bridge-Leg Controls the Midpoint-Voltage.

- **Back-to-back Connected IGBTs are Used.**
 - Using 1200V IGBTs-
- **Switches are Turned on only Twice a Period:**
 - Can be Optimized for Low Conduction Losses.
- **Four Step Commutation Sequence is Required.**

Switching Scheme



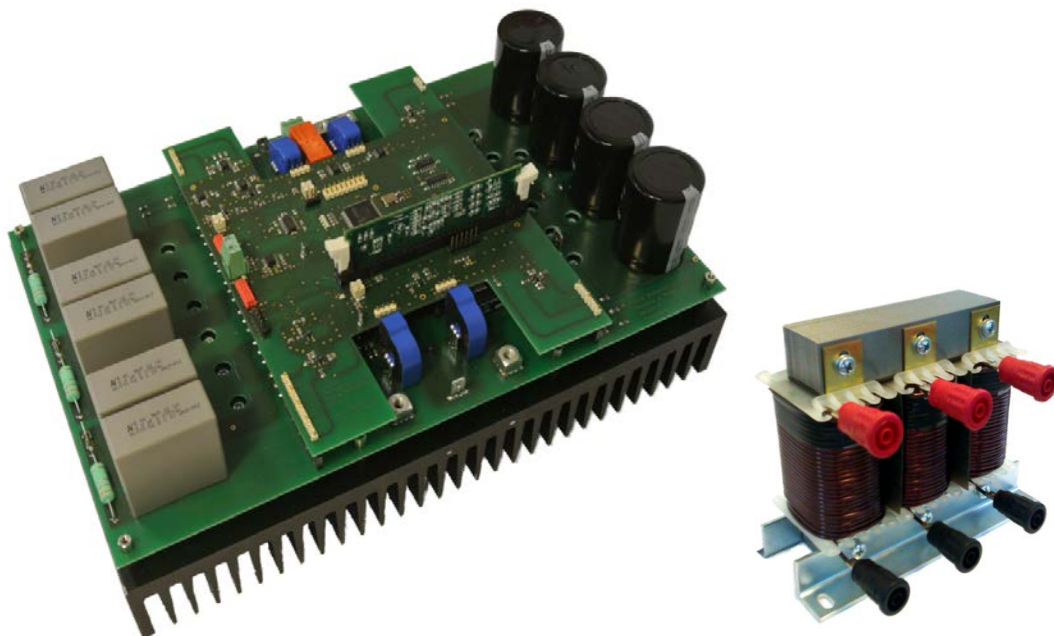
- **Switching Instants are Essential for High Input Current Quality.**
- **Analog Comparators in Combination with a CPLD.**
- **Complex Multi-Device Commutation:**
 - Rectifier Diodes
 - Bidirectional Switches
 - Filter Capacitors



- System Parameters:**

$V_{LL}=400$ V, $f_N=50$ Hz, $P_o=10$ kW, $I_{N,rms}=14.5$ A,
 $f_s=10$ kHz, $L_{h3}=L_{cp}=L_{cn}=3$ mH

- Average Midpoint Voltage is Controlled to Zero.**

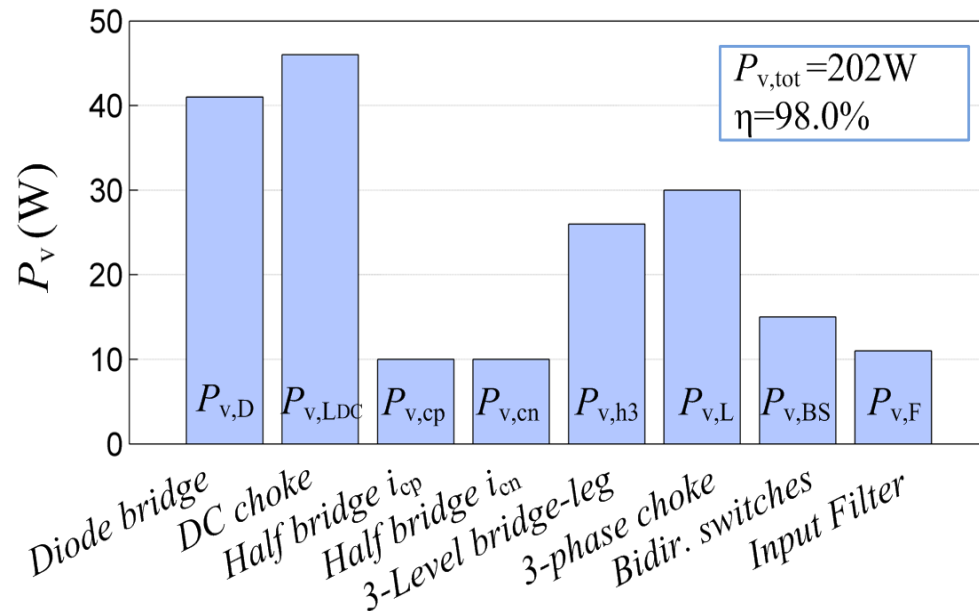


Specifications:

V_{LL}	400 V _{rms}
f_N	50/60 Hz
f_s	10 kHz
V_{cp}, V_{cn}	2 x 400 V _{DC}
P_o	10 kW
L_{cp}, L_{cn}, L_{h3}	2.6 mH

FCC: 300 mm x 200 mm X 97 mm

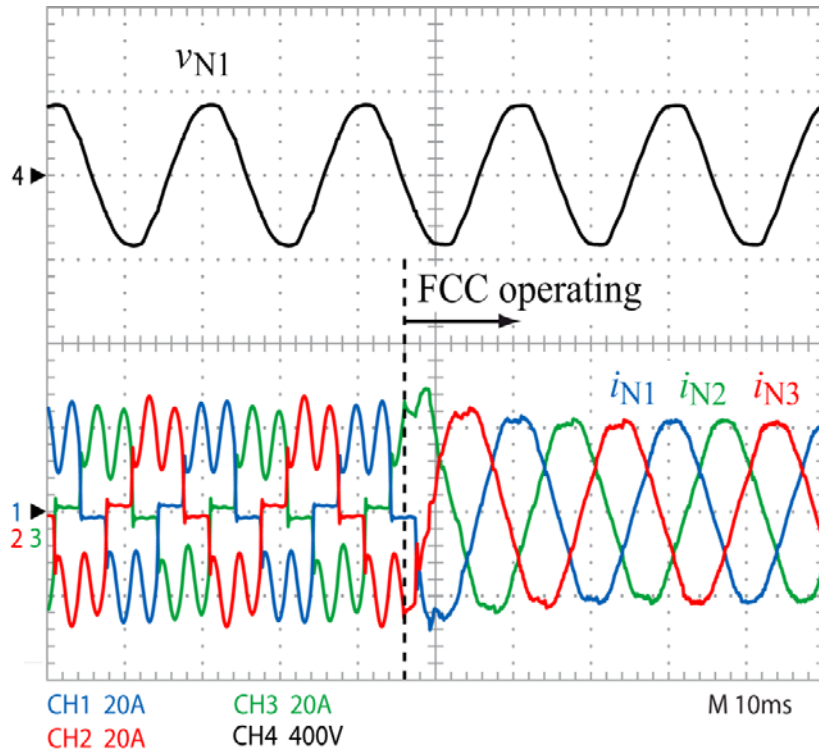
- **Scaled Demonstrator for Higher Power Ratings (e.g. 200 kW).**
 - Rather Small Switching Frequency of 10 kHz is Chosen.
 - Small Switching Frequency is a Challenge for Controllers.
- **Coupled Three-Phase Choke is Used.**
- **Prototype is not Optimized (Efficiency, Power Density, etc.)**



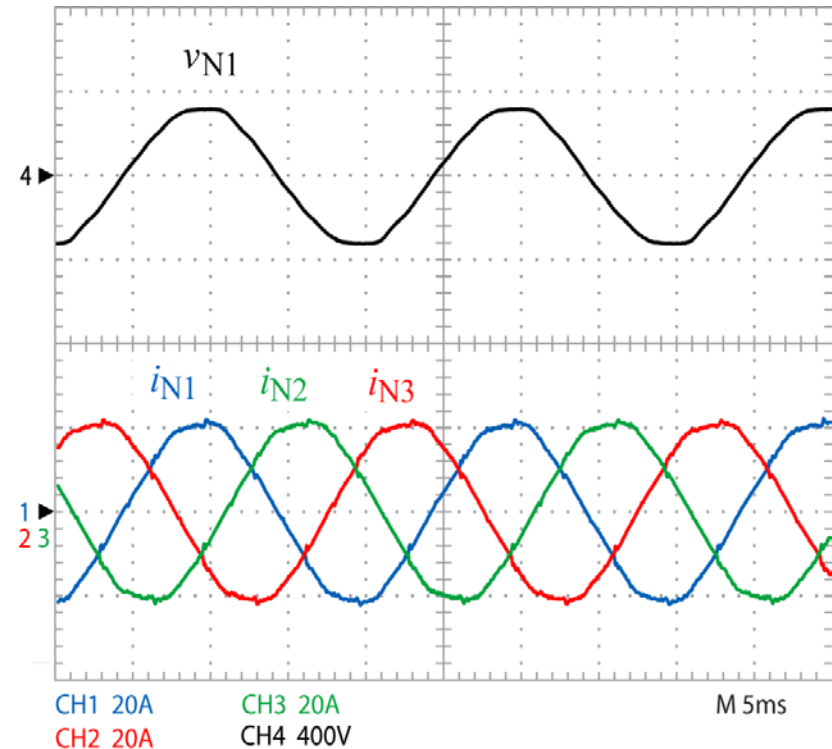
- **Losses are Dominated by Diode-Bridge and DC-choke (Handle 95% of P_o).**
- **FCC Losses are Mainly at 3-Level Bridge-Leg and 3-Phase Choke:**
 - **Primarily Switching Losses, Reduction by Application of SiC Diodes.**
 - **Reduction of Choke-Losses by Better Material in 3-Phase Choke.**

Transition

Diode Mode → to Active Current Shaping



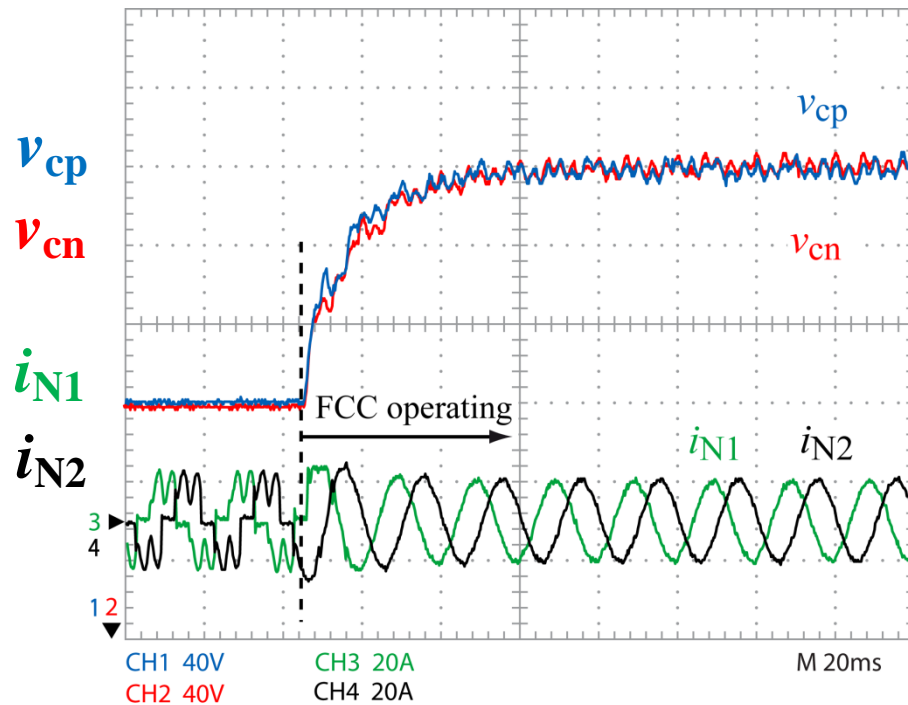
Continuous Operation



$$P_o=10\text{kW}, \text{THD}_i=2.3\%, \\ \lambda=0.998$$

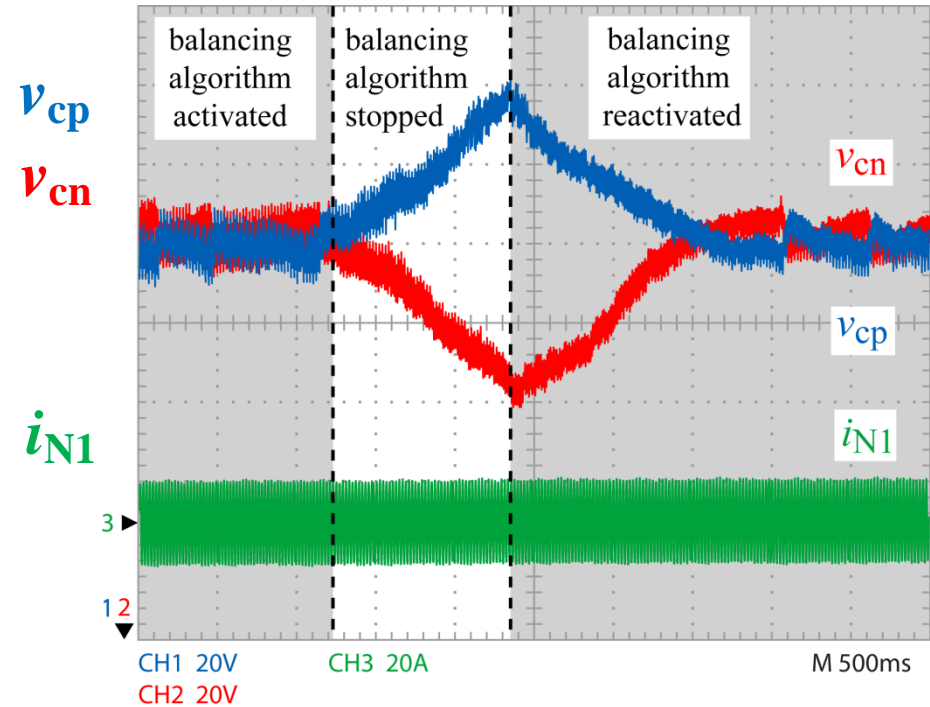
No Nameable Distortions in Current Shapes.

Startup of Operation



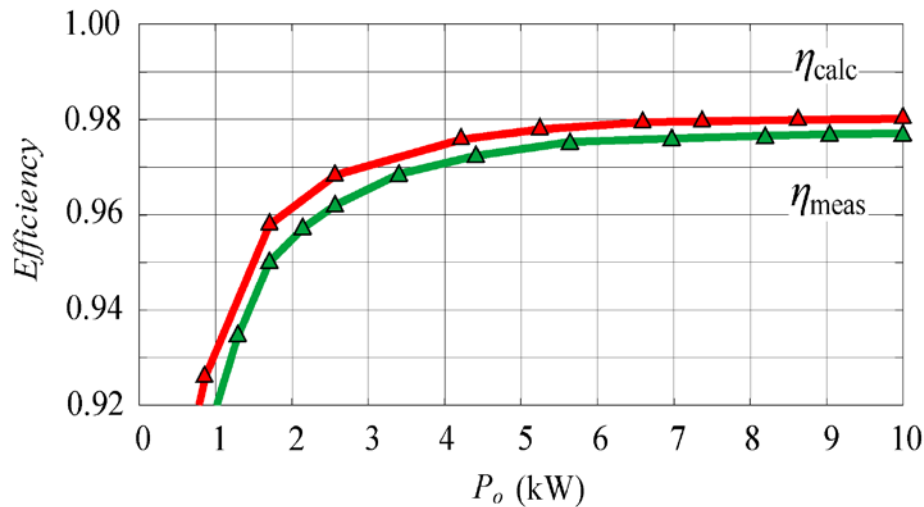
$P_o=10\text{kW}$, $V_{LL}=400\text{V}$,
 $f_N=50\text{Hz}$

Balancing



$P_o=10\text{kW}$, $V_{LL}=400\text{V}$,
 $f_N=50\text{Hz}$

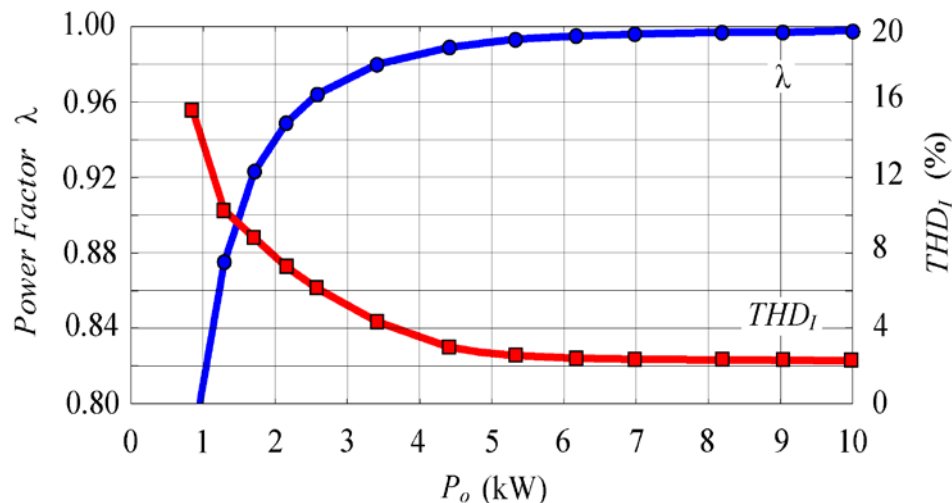
Measured Efficiency:



$$\eta_{meas} = 97.5\%$$

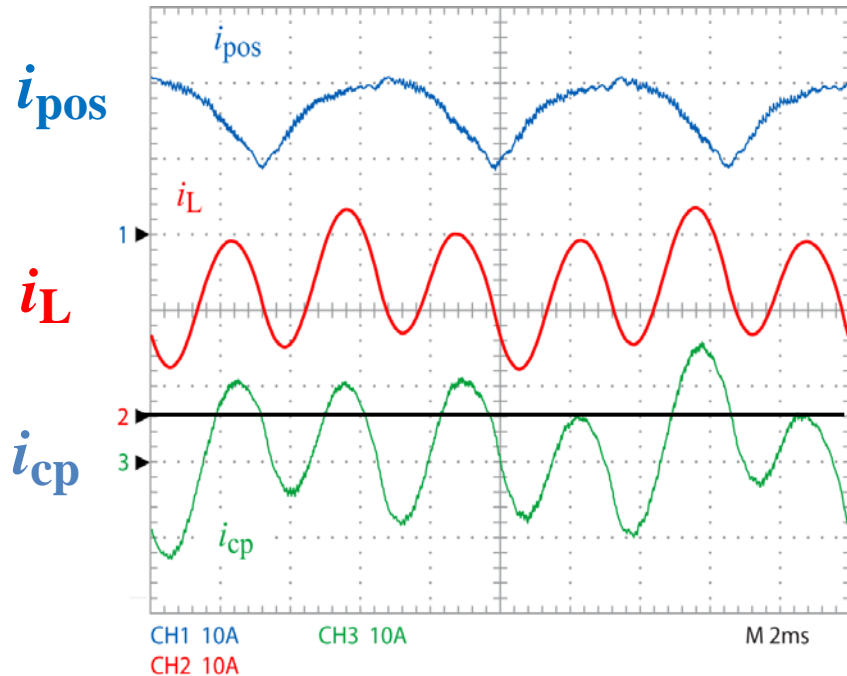
Measured Efficiency is Slightly Smaller:
– Higher Losses in 3-Phase Choke.

Measured Power Factor / Input Current Quality:

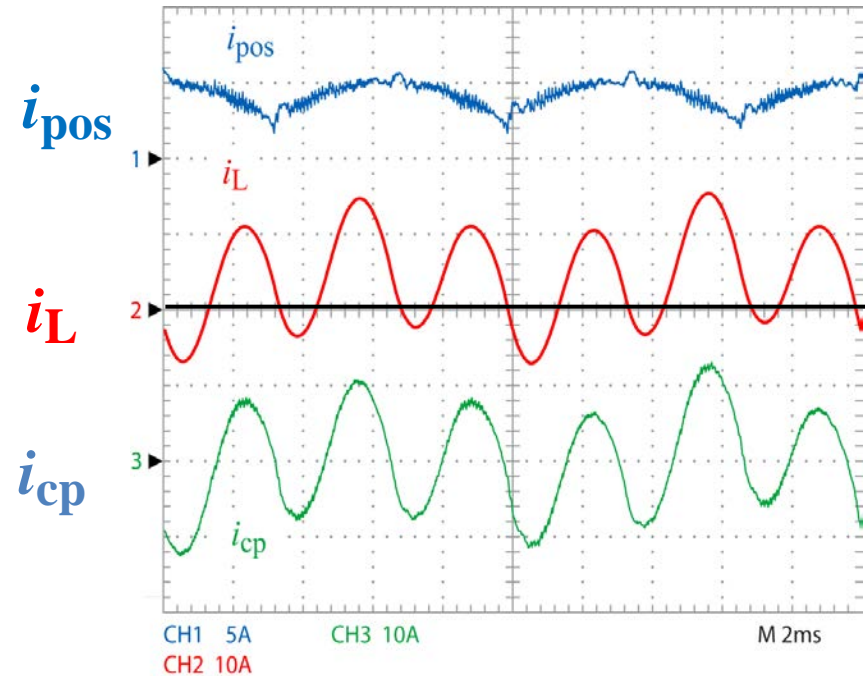


● **Very Good Input Current Quality:**
– Also at Partial Load.

Operating at $P_o = 10 \text{ kW}$



Operating at $P_o = 2.6 \text{ kW}$



- DC-choke Current gets Slightly Negative at Light-Load Condition.
- Results in a Circulating Current; Reduces Efficiency at Light Load.
- FCC can be Turned OFF at Light Load Condition.

- **Operating Principle of “Flying” Converter Cell is Verified:**
 - Allows Extension of Existing Diode Bridge to Low Harmonic Input Stage
 - Under Retention of the DC-Choke and the Large Output Capacitor.
 - Applications Where Controlled DC-Link Voltage is not Required.
- **Good System Performance:**
 - Processes Only a Small Amount of Output Power → Efficiency of **97.5%**.
 - No Intrinsic Input Current Distortion → Very Good THDi also at Light Load.
- **Suitable Control of Rectifier System is proposed**
 - Three controllers are required.
 - Implemented in a DSP and Small CPLD.
- **No High-Frequency Common Mode Voltage at the DC-Link.**

Thank you!
Questions?

